
通訊系統 (II)

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Chapter 3 Hybrid Amplitude/Phase Modulation

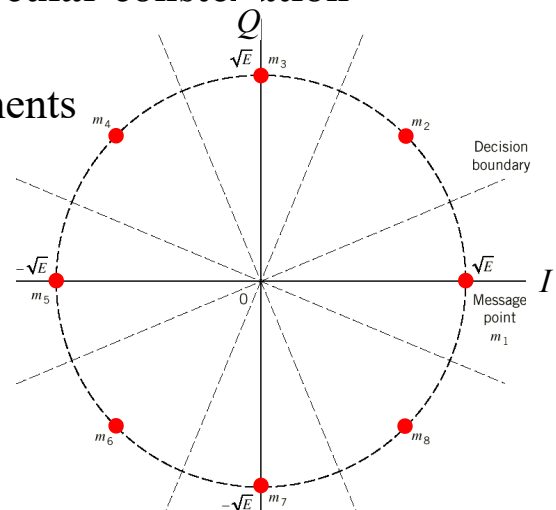
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Introduction

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Introduction

- In an M -ary PSK system, the **in-phase** and **quadrature** components are **inter-related**
 - The envelope is constrained to remain **constant**
 - The message points forms a **circular constellation**
- If the constraint is removed: the **in-phase** and **quadrature** components are permitted to be **independent**
 - A new modulation: M -ary **quadrature amplitude modulation (QAM)**
 - The carrier experiences **amplitude** as well as **phase** modulation



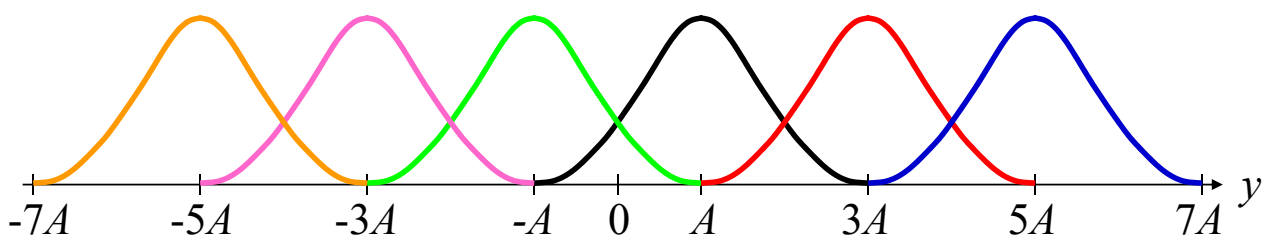
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M -ary Amplitude-Shift Keying (ASK)

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Baseband M -ary Pulse-Amplitude Modulation

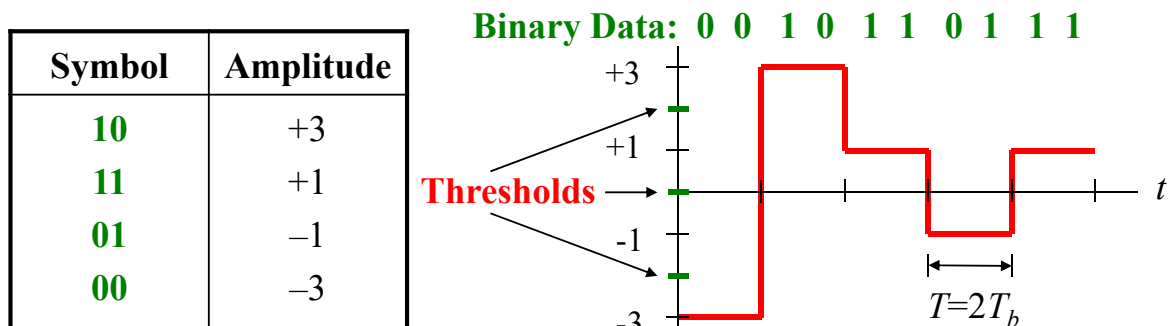
- In a **baseband M -ary Pulse-Amplitude Modulation (PAM)** system, the modulator produces one of M possible **amplitude levels** with $M > 2$
 - Each symbol contains **$\log_2 M$ bits** of data
 - A symbol is represented by a specified **amplitude level**
- To realize **the same average probability of symbol error**
 - The amplitude levels are $\pm A, \pm 3A, \pm 5A, \dots, \pm(M-1)A$
 - The maximum amplitude level is **about MA** for $M \gg 2$



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Bandpass Amplitude-Shift Keying (ASK)

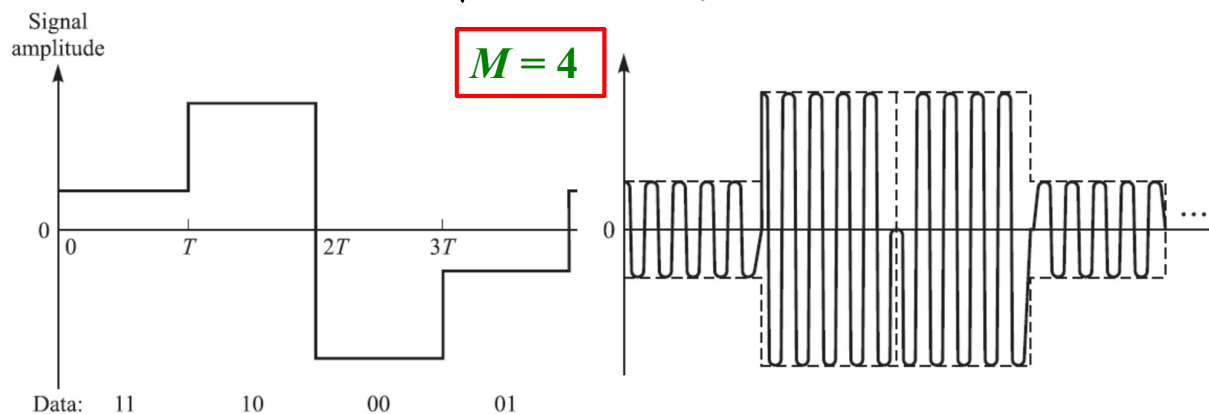
- The **bandpass** digital M -ary PAM is also called M -ary **Amplitude-Shift Keying (ASK)**.
- The mapping or assignment of $k = \log_2 M$ information bits to the $M = 2^k$ possible signal amplitudes can be done in many ways
 - In demodulation, the most likely errors caused by noise involve the erroneous selection of an **adjacent** amplitude
 - Gray encode**: the adjacent signal amplitudes differ by one bit



Signal Space of M -ary ASK

- In the **on-off keying (OOK)** version of an ASK system
 - Symbol 1 is represented by transmitting a sinusoidal carrier
 - Symbol 0 is represented by switching off the carrier
- There is only **one basis function** of unit energy

$$\phi_1(t) = \sqrt{2/T} \cos(2\pi f_c t), 0 \leq t < T$$

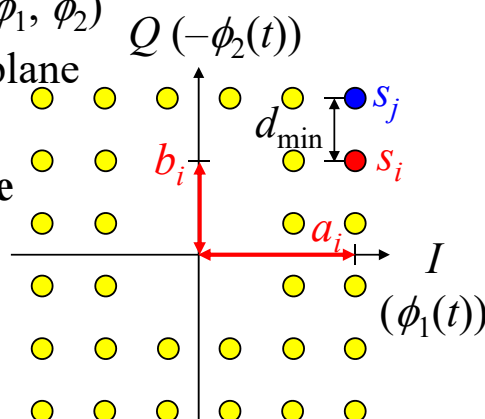


M-ary Quadrature Amplitude Modulation (QAM)

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M-ary Quadrature Amplitude Modulation(QAM)

- **M-ary Quadrature Amplitude Modulation (QAM)** is a **two-dimensional** generalization of M-ary PAM
- The two **orthonormal basis functions** are
 $\phi_1(t) = \sqrt{2/T} \cos(2\pi f_c t), 0 \leq t < T; \quad \phi_2(t) = \sqrt{2/T} \sin(2\pi f_c t), 0 \leq t < T$
 - A symbol is mapped to a signal point on the **two-dimensional** plane constructed by (ϕ_1, ϕ_2)
- Let the message point \mathbf{s}_i in the (ϕ_1, ϕ_2) plane be denoted by $(a_i d_{\min}/2, b_i d_{\min}/2)$
 - where d_{\min} is the **minimum distance** between any two message points
 - a_i and b_i are **integers**, for $i = 1, 2, \dots, M$



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Signal Space of M -ary QAM

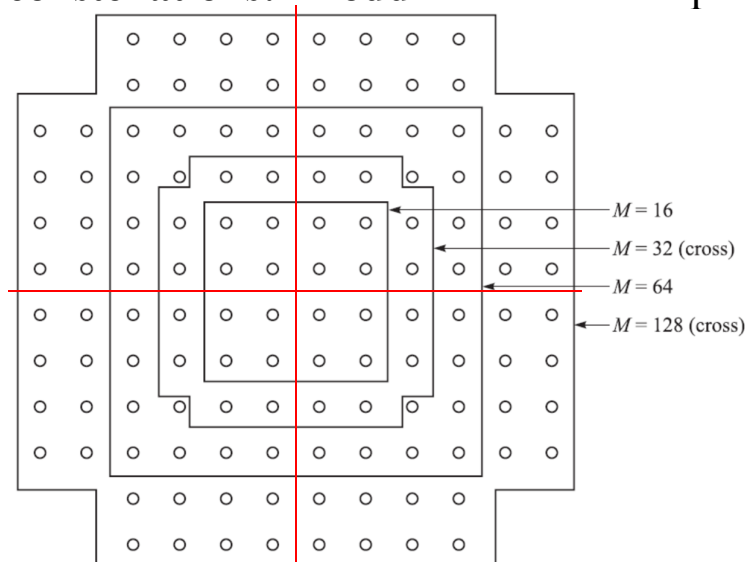
- Let $d_{\min} = 2\sqrt{E_0}$, where E_0 is the **one-dimension energy** of the signals with the **lowest amplitude** ($a_i = \pm 1$ or $b_i = \pm 1$)
- The transmitted M -ary QAM signals are defined by

$$s_k(t) = \sqrt{\frac{2E_0}{T}} \underbrace{a_k}_{\text{blue}} \underbrace{\cos(2\pi f_c t)}_{\text{red}} - \sqrt{\frac{2E_0}{T}} \underbrace{b_k}_{\text{blue}} \underbrace{\sin(2\pi f_c t)}_{\text{red}}, \quad 0 \leq t < T$$

- **Two phase-quadrature carriers:** $\cos(2\pi f_c t)$ and $\sin(2\pi f_c t)$
 - **Amplitude modulation:** a_k and b_k
- \Rightarrow **Quadrature Amplitude Modulation**

M -ary QAM Constellations

- There are two distinct QAM constellations:
 - **Square constellations:** an **even** number of bits per symbol
 - **Cross constellations:** an **odd** number of bits per symbol



M-ary QAM Signals: Square Constellations

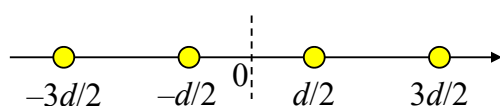
- With an even number of bits per symbol, we have $L = \sqrt{M}$
- An M -ary QAM square constellation can always be viewed as the **Cartesian product** of a **one-dimensional L -ary PAM constellation** with itself
 - The set of **all possible ordered pairs of coordinates**
 - The **first** (**second**) coordinate is taken from the **first** (**second**) set

First set **Second set**

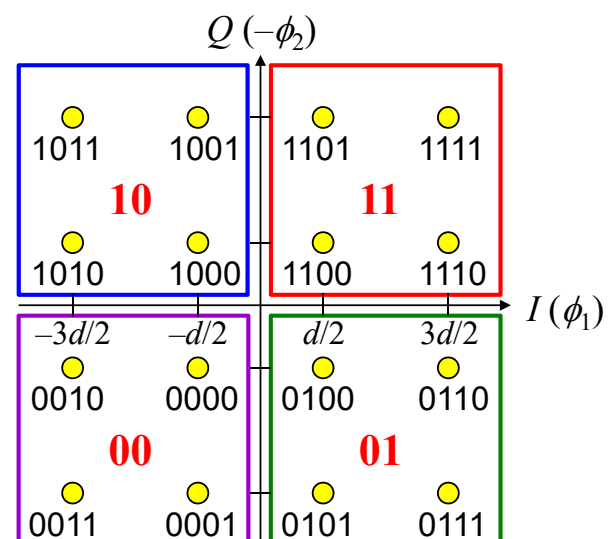
$$\{a_i, b_i\} = \begin{bmatrix} (-L+1, L-1) & (-L+3, L-1) & \cdots & (L-1, L-1) \\ (-L+1, L-3) & (-L+3, L-3) & \cdots & (L-1, L-3) \\ \vdots & \vdots & & \vdots \\ (-L+1, -L+1) & (-L+3, -L+1) & \cdots & (L-1, -L+1) \end{bmatrix}$$

Example 3

- Consider a 16-QAM constellation: 4 bits per symbol
 - The **left-most two bits**: a **quadrant** in the $\{\phi_1, \phi_2\}$ -plane
 - The **right-most two bits**: one **signal point** in a quadrant
 - The data mapping of signal points still follows the **Gray coding rule**
 - Any two nearest neighbors differ in only **one bit**



**One-dimensional
4-ary PAM constellation**



Example 3 (Cont.)

- The matrix of the two-dimensional coordinate is

$$\{a_i, b_i\} = \begin{bmatrix} (-3, +3) & (-1, +3) & (+1, +3) & (+3, +3) \\ (-3, +1) & (-1, +1) & (+1, +1) & (+3, +1) \\ (-3, -1) & (-1, -1) & (+1, -1) & (+3, -1) \\ (-3, -3) & (-1, -3) & (+1, -3) & (+3, -3) \end{bmatrix}$$

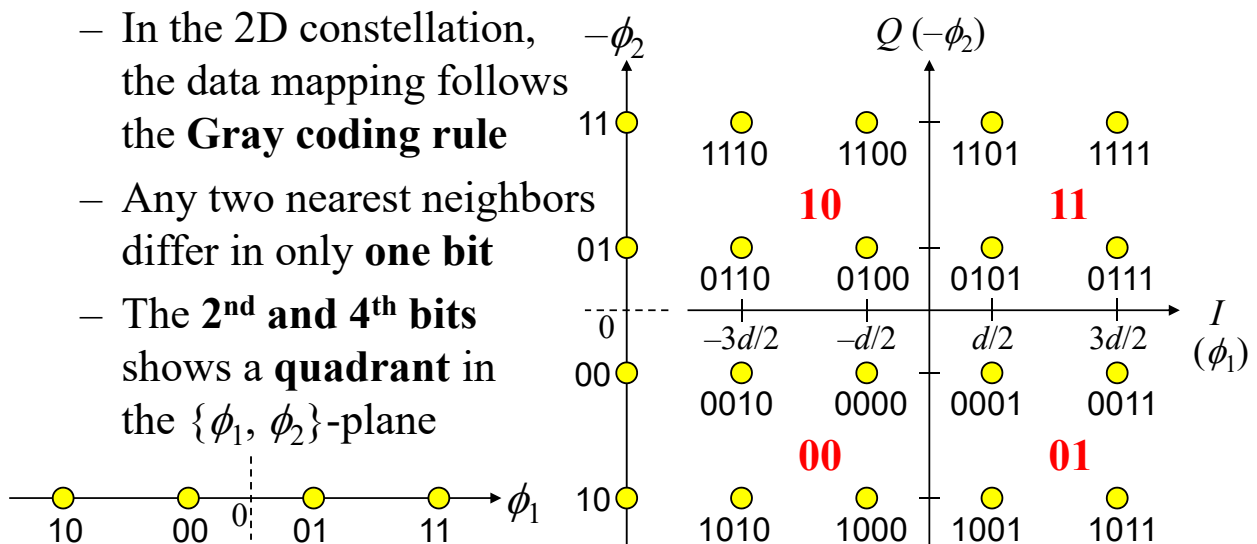
Another Square M -ary QAM Data Mapping

- Consider a 16-QAM constellation: 4 bits per symbol
- The data mapping in each of the **one-dimensional** 4-ary PAM constellation (ϕ_1 or ϕ_2) follows the **Gray coding rule**

- In the 2D constellation, the data mapping follows the **Gray coding rule**

- Any two nearest neighbors differ in only **one bit**

- The **2nd and 4th bits** shows a **quadrant** in the $\{\phi_1, \phi_2\}$ -plane



Error Probability of M -ary QAM: Square

- A QAM square constellation can be factored into the product of two PAM constellations

- The **probability of correct detection** for M -ary QAM is

$$P_c = (1 - P_e')^2$$

- where P_e' is the **probability of symbol error** for the corresponding L -ary PAM constellation with $L = \sqrt{M}$

$$P_e' = \left(1 - \frac{1}{\sqrt{M}}\right) \text{erfc}\left(\sqrt{E_0/N_0}\right), \quad E_0 = (d_{\min}/2)^2$$

- The **probability of symbol error** for M -ary QAM is

$$P_e = 1 - P_c = 1 - (1 - P_e')^2 \approx 2P_e'$$

- where it is assumed that P_e' is small enough

- Hence,

$$P_e \approx 2 \left(1 - \frac{1}{\sqrt{M}}\right) \text{erfc}\left(\sqrt{\frac{E_0}{N_0}}\right)$$

Error Probability of M -ary QAM: Square (Cont.)

- The **probability of symbol error** for the corresponding L -ary PAM constellation with $L = \sqrt{M}$:

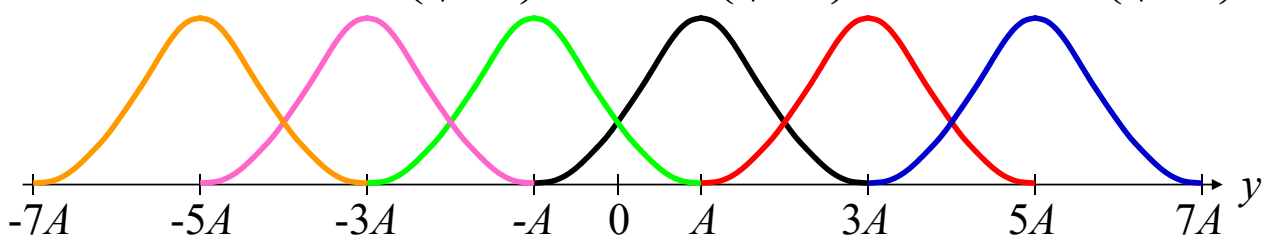
- For binary PAM: One-sided error $P_e' = \frac{1}{2} \text{erfc}\left(\sqrt{\frac{E_0}{N_0}}\right)$

- For L -ary PAM:

- Two-sided error: $(L - 2)$ points

- One-sided error: 2 points

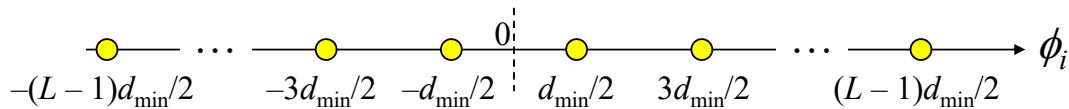
$$P_e' = \frac{2 \times (L - 2)}{2L} \text{erfc}\left(\sqrt{\frac{E_0}{N_0}}\right) + \frac{2}{2L} \text{erfc}\left(\sqrt{\frac{E_0}{N_0}}\right) = \left(1 - \frac{1}{L}\right) \text{erfc}\left(\sqrt{\frac{E_0}{N_0}}\right)$$



Symbol Energy of M -ary QAM: Square

- The transmitted symbol energy in M -ary QAM is variable
 - Depending on the transmitted **data** (the **signal point**)
 - For example, in the 16-QAM constellation
 - $2E_0$ (E_0+E_0): 4 signal points
 - $10E_0$ ($9E_0+E_0$): 8 signal points
 - $18E_0$ ($9E_0+9E_0$): 4 signal points
- The average symbol energy in 1D L -ary PAM constellation is
 - Assuming that the L levels are equiprobable

$$E'_{av} = \frac{2}{L} \times \left[E_0 + 9E_0 + 25E_0 + \dots + (L-1)^2 E_0 \right] = \frac{2E_0}{L} \sum_{i=1}^{L/2} (2i-1)^2$$



Symbol Energy of M -ary QAM: Square (Cont.)

- The average symbol energy in the M -ary QAM constellation is
 - The sum of the energy of the two PAM constellations

$$\begin{aligned} E_{av} &= 2 \times E'_{av} = \frac{4E_0}{L} \sum_{i=1}^{L/2} (2i-1)^2 \\ &= \frac{2(L^2-1)E_0}{3} = \frac{2(M-1)E_0}{3} \end{aligned}$$

- The **probability of symbol error** for M -ary QAM can be expressed as a function of E_{av}

$$P_e \approx 2 \left(1 - \frac{1}{\sqrt{M}} \right) \text{erfc} \left(\sqrt{\frac{3E_{av}}{2(M-1)N_0}} \right)$$

- For $M=4$ (i.e., QPSK)

$$P_e \approx \text{erfc} \left(\sqrt{E_{av}/2N_0} \right) = \text{erfc} \left(\sqrt{E/2N_0} \right)$$

M-ary QAM Signals: Cross Constellations

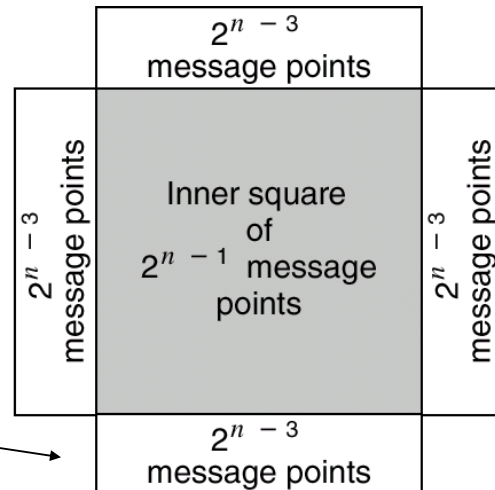
- When there is an odd number of bits per symbol (n) in M -ary QAM, a **cross constellation** is used
 - Start with a **square** constellation with $n - 1$ bits per symbol
 - Extend **each side** of the square constellation by **adding** 2^{n-3} signal points
 - Ignore the **corners** in the extension

$$2^n = 2^{n-1} + 4 \times 2^{n-3} = 2^{n-1} + 2^{n-1}$$

$$2^{n-3} = 2^{(n-1)/2} \times 2^{(n-5)/2}$$

- It is **not possible** to perfectly Gray code a cross constellation

Ignore to reduce average symbol energy



Error Probability of M -ary QAM: Cross

- Unlike the square constellations, a **cross constellation cannot** be expressed as the product of two 1D PAM constellations
- The determination of the symbol error probability for M -ary QAM with a cross constellation is complicated
- The **symbol error probability** for a cross constellation can be approximated as the result for a square constellation
 - Approximated as a square constellation with $M' = 2M$

$$P_e \approx 2 \left(1 - \frac{1}{\sqrt{2M}} \right) \text{erfc} \left(\sqrt{\frac{E_0}{N_0}} \right), \quad \text{Cross Constellation}$$

$$P_e \approx 2 \left(1 - \frac{1}{\sqrt{M}} \right) \text{erfc} \left(\sqrt{\frac{E_0}{N_0}} \right), \quad \text{Square Constellation}$$

$M' = 32$
 $M = 16$

Homework

- **You must give detailed derivations or explanations, otherwise you get no points.**
- Communication Systems, Simon Haykin (4th Ed.)
- 6.15;
- 6.16;